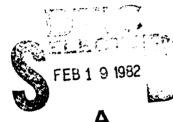
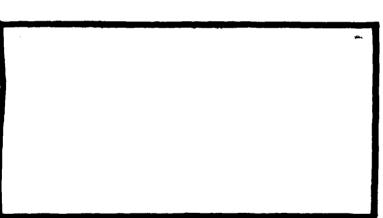


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A PERTURBATION ANALYSIS OF MODAL CONTROL OF AN UNSTABLE PERIODIC ORBIT

THESIS

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A PERTURBATION ANALYSIS OF MODAL CONTROL OF AN UNSTABLE PERIODIC ORBIT

THESIS

Presented to the Faculty of the School of Engineering
of the Air Force Institute of Technology
Air University
in Partial Fulfillment of the
Requirements for the Degree of
Master of Science

by

Dennis W. Ehrler

Captain USAF

Graduate Astronautical agineering

December 1981

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Acknowledgements

I would like to express my appreciation to Dr. William E. Wiesel for the encouragement, enthusiam and guidance that he provided me in this study. I owe Dr. Wiesel a special debt of gratitude for his constant help, advice and understanding.

I also would like to extend a special thanks to my wife. Lisa, for her understanding and encouragement throughout my course of study at the Air Force Institute of Technology.

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List of Symbols

| A | coefficient matrix of the variational system |
|----------------|---|
| B | control vector |
| CT. | control term |
| Ē | eigenvectors of the four-body system |
| е | exponentiation |
| F | partial deriotives of the Hamiltonian |
| G | universal gravitational constant |
| н | Hamiltonian of the four-body system |
| H.O.T. | higher order terms |
| I | identity matrix |
| J | Jordan form of the matrix of Poincare exponents |
| k | controller gain |
| L1-L5 | Lagrange points of equilibrium in earth-moon |
| | system |
| Λ | matrix of eigenvectors |
| λ | Poincare exponents of the four-body system |
| М | sum of the masses of the earth, moon, and sun |
| m ₁ | mass of the earth |
| ^m 2 | mass of the moon |
| m 3 | mass of the sun |
| m | mass of the satellite |
| n | modal vector of the four-body problem |

n p particular solution to the modal vector differential equation momenta of the satellite Px, Pv, Pz first order perturbation caused by the moon's orbital eccentricity state transition matrix coordinates of the satellite position qx,qy,qz vector from the earth-moon center of mass to the satellite r time derivative of R taken with respect to the rotating reference frame r vector from the earth to the moon \overline{r}_{sat} inertial position vector of the satellite vector from the earth-moon center of mass to the sun time derivative of p taken with respect to the rotating frame time initial time to kinetic energy of the satellite specific kinetic energy of the satellite period of the orbit about L3 control vector ν potential energy of the satellite

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velocity vector of the satellite

マュ

specific potential energy of the satellite

| ω | angular velocity vector of the rotating reference |
|------------------|---|
| | frame |
| × | state vector of the satellite |
| \overline{x}_h | homogeneous solution to the state vector differential |
| | equation |
| \overline{x}_p | particular solution to the state vector differential |
| | equation |
| ×n | solution state vector |
| δ 😾 | variational state vector |

Abstract

A Fourier representation of the perturbation effects of the moon's eccentricity on a satellite in orbit about L3 was constructed and verified numerically. It was then incorporated in a modal control scheme for linear systems that are time periodic which was developed by Shelton (Ref. 3). The feasibility of this modified control scheme was then verified by computer simulation by controlling the actual non-linear motions of the satellite with the modified controller.

A PERTURBATION ANALYSIS OF MODAL
CONTROL OF AN UNSTABLE PERIODIC ORBIT

Introduction

Background

Space is fast becoming a part of our everyday lives. Sensor-bearing satellites placed in orbit around the earth play a very important role in the world we live. From terrain and resource mapping to communication and strategic reconnaisance our lives are all affected by our exploits in space.

A desire to use the minimum number of satellites to satisfactorily accomplish mission objectives and maintain the orbits of these satellites has led to much research attempting to develop better ways to accomplish this task. A number of space operations require total coverage of the earth's surface by a particular set of satellites. Currently, such space operations are conducted by satellites positioned in earth-synchronous orbits, but not only does this not provide total global coverage, these orbits are unstable and require some sort of orbit maintaining device to be carried by the satellites. This control system, and the fuel to power it, must be carried as an integral part of the satellite, taking away, pound for pound, available payload weight allocated to

the sensors and support equipment required to perform the satellite's primary mission. Also, when the control system's fuel supply is exhausted, the satellites service life is at an end. These factors combine to force design tradeoffs between service life versus mission capability.

In recent years global strategic reconnaisance and communication have become increasingly important and, at the same time, more vulnerable to disruption. With the Soviet Union's increased research and development efforts aimed at developing "satellite killers" a less vulnerable position for these satellites is desirable.

The search for new orbits which would provide a maximum amount of, or total, earth coverage, hold down the control cost of maintaining these orbits and still decrease vulnerability to attack has led to extensive research into the "Lagrange" points. The five Lagrange points of equilibrium shown in Figure 1 represent theoretical points in space, in a coordinate system that rotates with the earth-moon system, where the combined gravitational effects of the earth and moon will keep a satellite placed at one of these points at that point. However, only L4 and L5 are stable and would provide long term stable orbits. Since the Lagrange points lie in the earth-moon plane and not in the earth's equitorial plane, one satellite placed at each of the points L2, L4, and L5 would provide "global" coverage, including the strategically valuable North Pole, at all times.

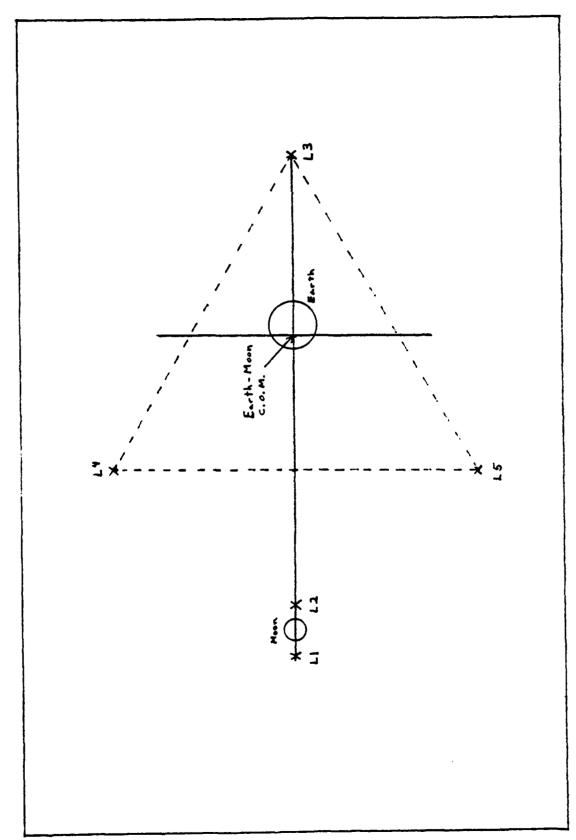


Figure 1. Lagrange Points in Earth-Moon System

Since the analysis from which the Lagrangian points are derived is based on the assumptions of circular, planar orbits and does not include the sun, when the "restricted four body problem" is analyzed all of the Lagrangian points become unstable. However, the points do provide a place to start.

Problem and Scope

Shelton developed a modal control scheme for linear systems that are time periodic. This scheme was then applied to a satellite in orbit about the earth-moon Lagrange point, L3. However, his results are only valid for a set of dynamics based upon his specific periodic orbit. The free lunar eccentricity, the lunar inclination, and the eccentricity of the sun were not included in his analysis.

Since the free lunar eccentricity is the major contributor of in-plane perturbations about the periodic orbit, this study will focus on its effects. With the active controller developed by Shelton operating, the periodic orbit becomes a stable reference solution and classical perturbation techniques can be applied to the addition of the free lunar eccentricity. If the solution, including the first order perturbation terms, is well behaved, free of resonances and unstable roots, then the controller will be modified to suppress only the "free" components causing the original instability. Higher order terms will hopefully be several orders of magnitude smaller than the first order terms

leaving us to believe if the controller operates satisfactorily in the lowest order perturbed system that the higher order terms are negligible. Applying this to Shelton's system we will again arrive at a control system with effectively zero long-term stationkeeping costs, but that is valid in the "real" representation of the dynamics and not just in the idealized system.

II. Problem Analysis

Modeling Assumptions

Shelton's research was based on a model of the three attracting bodies (earth, moon, and sun) that placed the moon in an orbit about the earth that is periodic but not circular. The earth orbit around the sun was modeled as circular and other inclination and eccentricity effects were initially ignored, restricting the motions of the four bodies to a plane. This study used these assumptions to rederive the Shelton controller followed by a perturbation analysis.

Coordinate System

Due to the nature of the periodicity of the orbit about L3, the coordinate system used in Shelton's research, and in this study, rotates with the average angular velocity of the moon and the center of the system is located at the earth-moon center of mass. This frame allows easy visualization of the periodic nature of the orbit about L3 (Ref. 5).

Derivation of the Hamiltonian (Ref. 6)

In deriving the dynamics for the restricted four-body problem, the three attracting bodies were assumed to have their actual masses while the satellite was assumed to be massless. This assures that the attracting bodies affect the motion of the satellite, but the satellite has no effect on

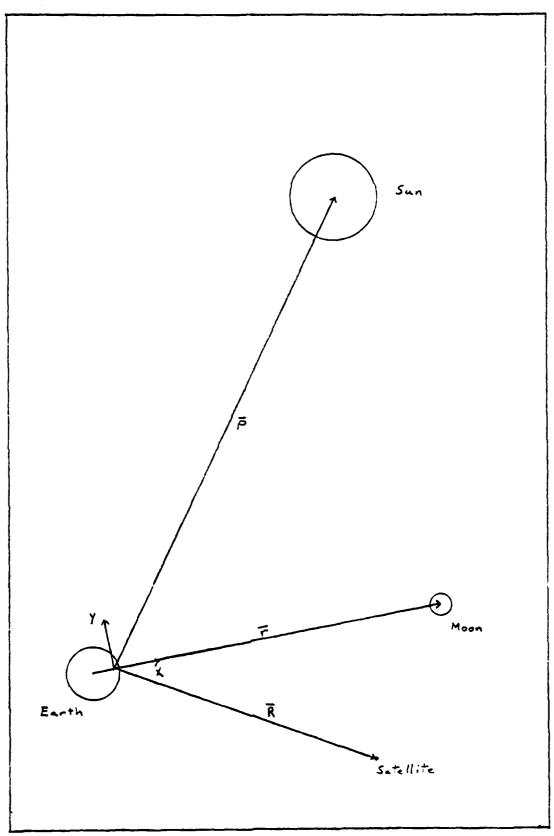


Figure 2. Geometry of the Four-Body Problem 7

the motion of the attracting bodies. The geometry of this system is depicted in Figure 2.

In order to derive the Hamiltonian the kinetic and potential energies must first be constructed. The inertial point in this derivation is assumed to be the earth-moon-sun center of mass. This results in the inertial position vector of the satellite to be

$$\overline{r}_{Sat} = \overline{R} - \frac{m_3}{M} \overline{\rho} \tag{1}$$

where $M = m_1 + m_2 + m_3$ (sum of the masses of the earth, moon, and sun, respectively) and the factor m_3/M which locates the inertial point of the system. Taking the inertial time derivative of the position vector in the rotating frame yields

$$\frac{i}{dt}(\overline{r}_{Sat}) = \overline{v}_{Sat} = \overline{R} + (\overline{\omega} \times \overline{R}) + (-\overline{M})^{\frac{m_3}{p}} + \overline{\omega} \times (-\overline{M})^{\frac{m_3}{p}}$$
(2)

where $-\frac{m_3}{M}$ $\frac{r}{\rho} + \frac{1}{\omega} \times (-\frac{m_3}{M})$ $\overline{\rho}$ is the transport velocity of the earth-moon center of mass. The symbol $\overline{\omega}$ denotes the angular velocity vector of the rotating frame, which is the moon's mean inertial angular velocity vector, and the superscript r represents differentiation in the rotating frame. The kinetic energy, T, of the satellite is defined as

$$T = 1/2 m (\overline{V}_{Sat} \cdot \overline{V}_{Sat})$$
 (3)

Because the satellite is assumed to be massless, the mass is divided from both sides which gives us the specific kinetic energy, T_s . Solving for T_s from equations (2) and (3) yields

$$T_{S} = 1/2 \left\{ \begin{array}{ccc} r \cdot & r \cdot \\ \overline{R} \cdot & \overline{R} + (\overline{\omega} \times \overline{R}) \cdot (\overline{\omega} \times \overline{R}) + \frac{-m_{3}}{M} & r \cdot \frac{-m_{3}}{M} & r \cdot \frac{-m_{3}}{M} \end{array} \right.$$

$$+ (\overline{\omega} \times \frac{-m_3}{M} \overline{\rho}) \cdot (\overline{\omega} \times \frac{-m_3}{M} \overline{\rho}) + 2 \Gamma R \cdot (\overline{\omega} \times \overline{R}) + 2 (\overline{R} \cdot \frac{-m_3}{M} \Gamma \cdot \overline{\rho})$$

$$+2\lceil \frac{-m_3}{R} \cdot (\overline{\omega} \times \frac{-m_3}{M} \overline{\rho}) \rceil + 2\lceil (\overline{\omega} \times \overline{R}) \cdot \frac{-m_3}{M} \overline{\rho} \rceil + 2\lceil (\overline{\omega} \times \overline{R}) \cdot (\overline{\omega} \times \frac{-m_3}{M} \overline{\rho}) \rceil$$

$$+2\left(\frac{m_3}{M}\right)^{\frac{1}{p}}\left(\overline{\omega}\times\frac{m_3}{M}\overline{\rho}\right)$$

The potential energy, V, of the satellite due to the gravitational attraction of the earth, moon, and sun is

$$V = \sum_{i=1}^{3} \frac{-Gm_{i}m}{r_{i}}$$
 (5)

where m is the mass of the satellite, r_i is the magnitude of the vector from the ith body to the satellite, and G is the universal gravitational constant. Dividing both sides of (5) by the mass of the satellite gives us the specific potential energy, V_s .

$$V_{S} = \frac{-Gm_{3}}{|\overline{R} - \rho|} - \frac{Gm_{2}}{|\overline{R} - \frac{m_{1}}{r}|} - \frac{Gm_{1}}{|\overline{R} + \frac{m_{2}}{r}|}$$
(6)

as the potential energy is not dependent on the time derivatives of the coordinates, the momenta, $\mathbf{p_i}$, of the Hamiltonian are

$$p_{i} = \frac{\partial T_{s}}{\partial \dot{q}_{i}} \tag{7}$$

where q_i is the time derivative of the ith coordinate. The coordinates used were the x, y, and z components of the satellite's position as depicted in Figure 2. In this frame the momenta, p_i , are the actual inertial velocity components, a fact that Shelton used to derive his controller.

The Hamiltonian is given by

$$H = \sum_{i=1}^{3} p_i q_i - T_s + V_s$$

$$(8)$$

Combining equations (4), (6) and (7) gives us the Hamiltonian.

$$H = \frac{1}{2}(p_{x}^{2} + p_{y}^{2} + p_{z}^{2}) + \omega (p_{x}R_{y} - p_{y}R_{x}) + \frac{m_{3}}{M}(r_{p_{x}} - \omega_{p_{y}})p_{x}$$

$$+ (\rho_{y} + \omega \rho_{x}) p_{y} - \frac{Gm_{3}}{|\overline{R} - \rho|} - \frac{Gm_{2}}{|\overline{R} - \frac{m_{1}}{m_{1} + m_{2}}} - \frac{Gm_{1}}{|\overline{R} + \frac{m_{2}}{m_{1} + m_{2}}}$$
(9)

From this Hamiltonian the equations of motion of the

satellite were formed.

Equations of Motion

The equations of motion for the satellite are derived from the Hamiltonian by forming

$$q_{m} = \frac{\partial H}{\partial p_{m}} \qquad m=1,2,\ldots,n$$
 (10)

$$\dot{p}_{m} = \frac{\partial H}{\partial q_{m}} \qquad m=1,2,\dots,n \tag{11}$$

where n represents the number of generalized coordinates describing the system (Ref. 5:94). These equations become

$$\frac{1}{x}(t) = \overline{f}(\overline{x}(t), t) \tag{12}$$

when put in state variable form. The state vector $\overline{\mathbf{x}}$ is made up of the coordinates and momenta, q's and p's, and $\overline{\mathbf{f}}$ contains the partial derivatives of the Hamiltonian defined by (10) and (11). In this restricted four-body problem the Hamiltonian is an explicit function of time so the equations of motion are explicit functions of time also.

Shelton assumed the existence of a periodic solution to (12), $\overline{x}_{c}(t)$. Defining

$$\delta \overline{x}(t) = \overline{x}(t) - \overline{x}_0(t)$$
 (13)

and differentiating (13) to get

$$\delta \frac{\dot{x}}{x}(t) = \frac{\dot{x}}{x}(t) - \frac{\dot{x}}{x}(t)$$
 (14)

Substituting (13) and (14) into (12) results in

$$\frac{\dot{x}}{x_0}(t) + \delta \dot{x}(t) = \bar{f}(\bar{x}_0(t) + \delta \bar{x}(t), t)$$
 (15)

Expanding this equation in a Taylor Series about $\overline{x}_0(t)$ produces

$$\frac{\dot{\overline{x}}}{(t)} + \delta \frac{\dot{\overline{x}}}{(t)} = \overline{f}(\overline{x}_{0}(t), t) + \frac{\partial \overline{f}}{\partial \overline{x}(t)} \Big|_{\delta \overline{x}(t)} + \text{H.O.T.}(16)$$

but from (12) we see that

$$\dot{\overline{x}}_0(t) = \overline{f}(\overline{x}_0(t), t) \tag{17}$$

which reduces (16) to

$$\delta \dot{\overline{x}}(t) = A(t) \delta \overline{x}(t) \tag{18}$$

including only first order terms, with

$$A(t) = \frac{\partial \overline{f}}{\partial \overline{x}(t)} | \overline{x}_{0}(t)$$
 (19)

Displacing the periodic solution , $\overline{x}_0(t)$, at t_0 an amount, $6\overline{x}$, and expanding the solution in a Taylor series about $x_0(t)$ yields

$$\overline{x}(t) = \overline{x}_0(t) + \frac{\partial \overline{x}(t)}{\partial \overline{x}(t_0)} \middle| \delta x(t_0) + H.O.T.$$
 (20)

Truncating to first order and using (13) produces

$$\delta \overline{x}(t) = \Phi(t, t_0) \delta \overline{x}(t_0)$$
 (21)

where

$$\Phi(t,t_0) = \frac{\partial \overline{x}(t)}{\partial \overline{x}(t_0)} \Big|_{\overline{x}_0(t)}$$
 (22)

The matrix, Φ , is called the state transition matrix and is a function of only the initial and final times being considered. The properties of the Φ matrix are

$$\phi(t,t_0) = \phi(t,t_1) \ \phi(t_1,t_0)$$
 (23)

and

$$\phi(t_0, t_0) = I \tag{24}$$

where I is the Identity matrix. Differentiating (21) with respect to time produces

$$\delta \dot{\overline{x}}(t) = \dot{\Phi}(t, t_0) \, \delta \overline{x}(t_0) \tag{25}$$

as $\delta \overline{x}(t_n)$ is a constant.

Floquet Theory

The four-body problem Hamiltonian, derived in a previous section, is an explicit function of time and not a constant of the motion. Energy is not conserved and no other constants of the motion exist (Ref. 5:416). The equations of motion that are derived from the Hamiltonian are also time varying. The variational equations of motion are written in state vector form, derived previously, as

$$\delta \dot{\overline{x}}(t) = A(t) \delta \overline{x}(t) \tag{26}$$

This system is periodic and the periodicity is contained in the coefficient matrix $A(\tau)$ where τ is defined as the period of the system. Differential equations such as (26), that have periodic coefficients, can be reduced to the case of constant coefficient, linear differential equations, by the theory of Floquet (Ref. 3:60). Floquet's theory states the general solution to (26) is of the form

$$\delta \overline{x}(\tau) = \sum_{j=1}^{n} \beta_{j} \overline{E}_{j}(t) e^{\lambda_{j} t}$$
(27)

where $\beta_{\dot{1}}$ and $\lambda_{\dot{1}}$ are constants and the $\overline{E}_{\dot{1}}$ are periodic func-

tions with the same period, τ . The n λ_j can be arranged in a matrix of the Jordan canonical form which basically diagonalizes the matrix (Ref. 5:267). The constant λ_j is called a characteristic or Poincare exponent of A(t).

In order to compute \overline{E}_j and λ_j Shelton assumed that he needed to excite one mode of the system described by (26). Letting β_j =0 for $j\neq i$ and β_i =1 yields, from (27),

$$\delta \overline{x}(t) = \overline{E}_{i}(t)e^{\lambda_{i}t}$$
 (28)

From (21), with $t_0=0$

$$\delta \overline{\mathbf{x}}(\mathbf{t}) = \Phi(\mathbf{t}, 0) \delta \overline{\mathbf{x}}(0) \tag{29}$$

and equating (28) and (29), yields

$$\overline{E}_{i}(t)e^{\lambda_{i}t} = \Phi(t,0)\delta \overline{x}(0)$$
 (30)

$$= \Phi(t,0) \overline{E}_{i}(0)$$
 (31)

Since $\overline{E}_i(t)$ is a periodic vector, $\overline{E}_i(\tau) = \overline{E}_i(t)$, results in

$$\delta \overline{\mathbf{x}}(\tau) = \overline{\mathbf{E}}_{\mathbf{i}}(\tau) \mathbf{e}^{\lambda} \mathbf{i}^{\tau} \tag{32}$$

$$= \overline{E}_{i}(0)e^{\lambda_{i}\tau} = \Phi(\tau, 0)\overline{E}_{i}(0)$$
 (33)

Rearranging (33) yields

$$[\Phi(\tau,0) - (e^{\lambda_i \tau}) I] \overline{E}_i(0) = \overline{0}$$
 (34)

which is an eigenvalue problem with the eigenvalues of $\Phi(\tau,0)$ being $e^{\lambda \tau}$. The eigenvectors of $\varphi(\tau,0)$ are, therefore, the vectors $\overline{E}_{\dot{1}}\left(0\right)$.

To complete the solution, \overline{E}_i (t) must be computed over one period. Since \overline{E}_i (t) is periodic this determines \overline{E}_i (t) for all time.

Combining equations (26) and (28) produces

$$\frac{d}{dt}(\overline{E}_{i}(t)e^{\lambda_{i}t}) = A(t)(\overline{E}_{i}(t)e^{\lambda_{i}t})$$
(35)

Defining $\Lambda(t)$ as

$$\Lambda(t) = [\overline{E}_{1}(t)] [\overline{E}_{2}(t)] \dots [\overline{E}_{n}(t)]$$
(36)

and incorporating Λ (t) into (35) yields

. Jt Jt Jt Jt
$$\Lambda(t)e + \Lambda(t)Je = \Lambda(t)\Lambda(t)e$$
 (37)

and reducing further gives the differential equation

$$\Lambda(t) + \Lambda(t)J = \Lambda(t)\Lambda(t)$$
 (38)

Shelton used the Floquet theory described above to investigate the stability of the orbit. His work demonstrated, as did work previously done by Wiesel and Smith,

that the orbit was unstable. One of the system's Poincare exponents is a positive real root, while its conjugate is negative and the others are purely imaginary.

Table I. Poincare Exponents of Shelton's Orbit

| Mode | Poincare Exponent |
|--------------|---------------------------------|
| Planar | 0.0 - 1.0392i 0.0 + 1.0392i |
| Planar | 2.3921 + 0.0i -2.3921 + 0.0i |
| Out of Plane | 0.0 - 1.1247i 0.0 + 1.1247i |

In Shelton's orbit, the unstable exponent was designated λ_3 and is the root that his controller affected (Ref. 6). This gave the following equation when the controller was incorporated

$$\delta \dot{\overline{x}}(t) = A(t) \delta \overline{x}(t) + B(t) \overline{u}(t)$$
 (39)

where $\overline{u}(t)$ is the control vector and B(t) is used to designate to which states the control will be applied. In order to design the control system so only the unstable mode is affected by the controller, a new variable was defined.

$$\overline{\eta}(t) = \Lambda (t) \overline{x}(t)$$
 (40)

where $\overline{\eta}(t)$ is the modal vector and $\Lambda(t)$ is as defined in (36). Taking the time derivative of equation (40) yields

$$\delta \dot{x}(t) = \dot{\Lambda}(t) \dot{\eta}(t) + \Lambda(t) \dot{\dot{\eta}}(t)$$
 (41)

Equating equations (39) and (41) and solving for $\dot{n}(t)$, to get a differential equation in terms of the modal variables, results in

$$\frac{\cdot}{\eta}(t) = \frac{-1}{\Lambda}(t) \lceil A(t) \Lambda(t) - \Lambda(t) \rceil \overline{\eta}(t) + \Lambda(t) B(t) \overline{u}(t)$$
 (42)

From (38)

$$\Lambda(t) = \Lambda(t) \Lambda(t) - \Lambda(t) J \qquad (43)$$

substituting (43) in (42) produces

$$\frac{\cdot}{\eta}(t) = J\overline{\eta}(t) + \Lambda (t)B(t)u(t)$$

where J is the constant diagonal matrix of Poincare exponents. Since the system is diagonalized, feeding back a single unstable mode can be accomplished, making it relatively simple to implement the control.

Shelton implemented this control scheme by selecting $\vec{u}(t) = U = kn_3$ where k is the constant control gain. This gave the following differential equation

The next chapter will discuss how, using Shelton's controlled system, a perturbation analysis was performed on the system and the controller was modified to ignore the oscillations forced on the orbit by the moon's eccentricity.

III. Perturbation Analysis

The solution of a differential equation of the sort

$$\frac{\cdot}{x} = \overline{A} x + \overline{F} \tag{45}$$

contains a free (homogeneous) and a forced (particular) solution such that the total solution is of the form

$$\overline{x} = \overline{x}_h + \overline{x}_p \tag{46}$$

with \overline{x}_p being the particular solution associated with a particular \overline{F} (forcing function). Shelton's controller is designed to suppress both the free and forced components of the unstable mode. The controller was designed for the idealized case with the moon in a circular orbit. When Shelton's controller is implemented with the moon in its real, slightly eccentric, orbit the controller does not keep the satellite's orbit stable. In order to incorporate the perturbations caused by the moon, in the variational equations the lowest order perturbation terms evaluated on the periodic orbit will be included such that

$$\frac{\cdot}{\delta \mathbf{x}} = \{ \mathbf{A}(\mathbf{t}) + \overline{\mathbf{B}} \mathbf{K} \Lambda (\mathbf{t}) \} \delta \mathbf{x} + \overline{\mathbf{P}}(\mathbf{t})$$
 (47)

where $K = [0 \ 0 \ k \ 0 \]$ when only n_3 is fed back, $\overline{P}(t)$ represents the perturbing effects of the moon's eccen-

-1 tricity and KA (t) \overline{x} is the Shelton controller. Defining

$$A'(t) = A(t) + \overline{B}K\Lambda (t)$$
 (28)

and applying the Floquet transform to the modal variables for this controlled system (see section on Floquet theory), we find

$$\frac{\cdot}{n}(t) = J'\overline{n}(t) + \Lambda'(t) \overline{P}(t)$$
 (49)

where the primes refer to the closed loop system, for a given gain k. This reduces the system (47) to constant coefficient form (Ref. 9).

The solution to the constant coefficient, linear differential equation is of the form (46). The forcing function

P(t) was developed, to lowest order, using

$$\overline{P}(t) = \frac{\dot{x}}{x_R}(t) - \frac{\dot{x}}{x_{P0}}(t)$$
 (50)

where $\frac{\cdot}{x_R}(t)$ is the derivative of the solution to the variational equation with the moon's actual eccentricity

included and $\frac{\cdot}{x}_{P0}(t)$ is the derivative of the solution with the moon in its idealized, circular orbit.

 $\frac{-1}{\Lambda'}\frac{1}{P(t)}$ was obtained by evaluating (50) at thirty equally spaced time steps around the orbit while holding the

angle associated with the phase of the moon constant,

multiplying times Λ ' (t), and then constructing a Fourier series to represent the elements at each time step. Keeping only the first N terms of the series yields

$$\frac{-1}{\Lambda'(t)} \overline{P}(t) = \sum_{i=0}^{N-1} \overline{\Lambda'_{i}} \cos(a_{i}) + \sum_{i=0}^{N-1} \overline{B'_{i}} \sin(a_{i})$$
(51)

where

$$a_i = i w_0 t$$

This procedure was repeated until the Fourier series for a number of small, constant step increments of the phase of the moon from 0 to 2π radians were obtained. Another Fourier series was then constructed from the coefficients, $A_{\hat{i}}$'s and $B_{\hat{i}}$'s, of these Fourier series. The result was a double sum Fourier series represented by

+
$$\overline{C}_{nm} \sin(a_n) \cos(b_m) + \overline{D} \sin(a_n) \sin(b_m)$$
 (52)

where the coefficients A_{nm} , B_{nm} , C_{nm} , and D_{nm} are the coefficients resulting from the construction of the overall Fourier series (52). The terms a_n and b_m are associated with the phase of the orbit and the phase of the moon, respectively, and are defined as

$$a_n = n w_0 t \tag{53}$$

$$b_{m} = m(w_{E}t + \Omega_{10})$$
 (54)

where n and m are the summation indices, \mathbf{w}_0 is the periodic orbit frequency, $\mathbf{w}_{\rm E}$ is the frequency of the line of apsides of the lunar orbit, and \mathbf{Q}_{10} is a constant phase.

Since the product of two trigonometric functions can also be represented as the sum and difference of two trig functions the following relationships were used to expand

-1 $\Lambda'(t) \overline{P}(t)$.

$$\sin(a)\sin(b) = \frac{1}{-\cos(a-b)} - \frac{1}{-\cos(a+b)}$$
 (55)

$$\cos(a)\cos(b) \approx \frac{1}{-\cos(a-b)} + \frac{1}{-\cos(a+b)}$$
 (56)

$$\sin(a)\cos(b) = \frac{1}{2}\sin(a+b) + \frac{1}{2}\sin(a-b)$$
 (57)

$$\cos(a)\sin(b) = \frac{1}{-}\sin(a+b) + \frac{1}{-}\sin(a-b)$$
 (58)

Defining

$$\phi_{nm} = a + b = (nw_0 + mw_E)t + mQ_{10}$$
 (59)

$$\psi_{nm} = a - b = (nw_0 - mw_E)t - mQ_{10}$$
 (60)

and substituting (55) through (60) into (52) produces

$$+(\overline{B}_{nm}-\overline{D}_{nm})\sin(\psi_{nm})+(\overline{B}_{nm}+\overline{D}_{nm})\sin(\phi_{nm})] \qquad (61)$$

From (49) we see that

$$\frac{\cdot}{\overline{\eta}^*}(t) = J^*\overline{\eta}^*(t) + \Lambda^*(t)\overline{P}(t)$$
 (62)

Since J' is of the form (Ref. 5)

$$J' = \begin{bmatrix} 0 & \alpha_1 & 0 & 0 \\ \alpha_2 & 0 & 0 & 0 \\ 0 & 0 & \alpha_3 & 0 \\ 0 & 0 & 0 & \alpha_d \end{bmatrix}$$
 (63)

where the α 's are the poincare exponents of the controlled system. Since $n_3(t)$ is used by the controller solving (62) for $n_{3p}(t)$ results in

$$n_{3p}(t) = \alpha_{3}n_{3p} + \lceil \Lambda'(t)\overline{P}(t) \rceil_{3}$$
(64)

Assuming a solution of the same type as the forcing function and taking its derivative we obtain

$$\eta_{3p}(t) = \sum_{L=1}^{\Gamma} {W_{L}\cos(\phi_{L}) + V_{L}\sin(\phi_{L}) + Y_{L}\cos(\psi_{L}) + Z_{L}\sin(\psi_{L})}$$
(65)

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$$-Y_{L}\psi_{L}\sin(\psi_{L}) + Z_{L}\psi_{L}\cos(\psi_{L})]$$
(66)

where W_L , V_L , Y_L , and Z_L are the coefficients we are solving for and L is related to the nm designator in (61).

Substituting (61), (65) and (66) into (64) and equating like terms yields (L corresponding to a particular nm, L = Mn + m + 1)

$$\sin(\phi_L): -W_L \dot{\phi}_L = \alpha_3 V_L + 1/2 (B_{nm} + D_{nm})$$
 (67)

$$\cos(\phi_L): V_L \dot{\phi}_L = \alpha_3 W_L + 1/2 (A_{nm} - C_{nm})$$
 (68)

$$\sin(\psi_L): -Y_L \dot{\psi}_L = \alpha_3 Z_L + 1/2(B_{nm} - P_{nm})$$
 (69)

$$\cos(\Psi_{L}): Z_{L}^{\dot{\Psi}}_{L} = \alpha_{3}Y_{L} + 1/2(A_{nm} + C_{nm})$$
 (70)

Solving (67) through (70) for W_L , V_L , Y_L and Z_L and substituting in

$$\phi_{L} = nw_{0} + mw_{E} \tag{71}$$

$$\dot{\psi}_{L} = n w_{0} - m w_{E} \tag{72}$$

we obtain the coefficients for the particular solution of $\eta_{\mathfrak{I}}(t)$ as follows

$$W_{L} = \frac{-\alpha_{3}V_{L} - 1/2(B_{nm} + D_{nm})}{(nw_{0} + mw_{E})}$$
(73)

$$V_{L} = \frac{\alpha_{3}W_{L} + 1/2(A_{nm} - C_{nm})}{(nw_{0} + mw_{E})}$$
 (74)

$$Y_{L} = \frac{-\alpha_{3}Z_{L}^{-1/2(B_{nm}-D_{nm})}}{(nw_{0} - mw_{E})}$$
 (75)

$$Z_{L} = \frac{\alpha_{3}Y_{L} + 1/2(A_{nm} + C_{nm})}{(nw_{C} - mw_{E})}$$
(76)

Numerically evaluating (73) through (76) and using them with (65) gives us a solution to $n_{3p}(t)$. To insure that this solution was correct $n_{3p}(t_1)$ was numerically calculated and using the approximation

$$n_{3p}(t_1) = \frac{n_{3p}(t_1) - n_{3p}(t_1 + \Delta t)}{\Delta t}$$
 (77)

and substituting these values into

$$P_3(t_1) = \Lambda(t_1) \ln_{3p}(t_1) + \alpha_{3n_{3p}}(t_1)$$
 (78)

This calculated value was compared with the value for $P_3(t)$ calculated from (50). The numbers matched to six decimal places using a $\Delta t = 10^{-6}$ sec. This makes me resonably sure that the correct solution for $n_{3p}(t)$ was formulated. The

derivations of $\eta_{1p}(t)$, $\eta_{2p}(t)$ and $\eta_{\Delta p}(t)$ are contained in Appendix A.

IV. Implementation of the Modified Controller

Having solved for the particular solution due to the perturbation effects of the moon Shelton's controller was modified to incorporate this particular solution. Shelton's controller stabilizes the idealized orbit by zeroing out the difference between the satellite's true orbit and the stable controlled periodic orbit. When the moon's perturbation effects are added in, the controlled orbit is no longer stabilized. The Fourier analysis of the perturbing force showed it to be purely oscillatory, with no secular terms, to first order. If we could calculate these terms and subtract them from the controlled term we could stabilize the orbit with the moon's proper dynamics involved.

The Shelton controller stabilizes the idealized system by calculating the control term given in (44)

$$CT. = k\Lambda (t)Bn_3(t)$$
 (79)

which, with the moon in its actual orbit, yield

$$CT. = k\Lambda (t)B[n_{3H}(t) + n_{3p}(t)]$$
 (80)

driving the system unstable. Having solved for the particular solution (65) the controller was implemented by subtracting off $\eta_{3p}(t)$ which yields

$$CT. = k\Lambda (t)B_{\eta_{3H}}(t)$$
 (P1)

This system produces a new stabilized orbit which is not the idealized periodic orbit that Shelton derived.

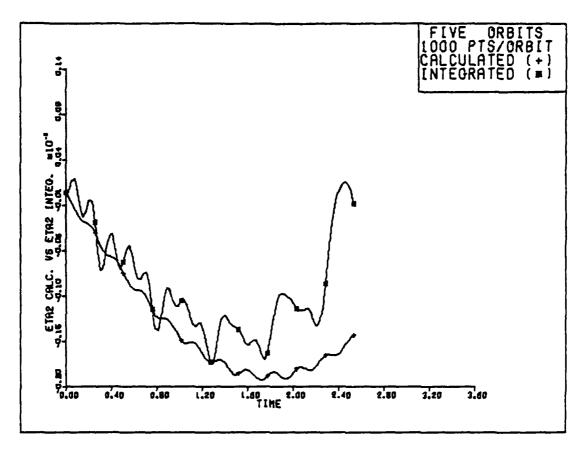
V. Results and Conclusions

Evaluation of the Controller

As stated in chapter IV of this report, the modified controller design was based on a linearized set of equations including a first order approximation of the forcing function caused by the eccentricity of the moon's orbit. To evaluate the controller both the actual integrated solution with the modified controller operating and the particular solution,

 $\overline{n}_p(t)$, were plotted to see if the integrated solution would track the calculated particular solution (Figure 3). All four modal variables were plotted to insure that the calculated solution was indeed tracking the integrated solution for each variable. As can be seen in Figure 3 the calculated and integrated solutions match very well.

During the sixth orbit the integrated solution becomes unstable (Figure 4). This occurs due to a combination of two factors, the gain selected for implementation of Shelton's controller, k=.459, was small and the particular solution was larger than expected. Because the particular solution was so much larger than expected, the error in the calculated term is also quite appreciable. This is because the perturbation effects were assumed to be first order and were evaluated with respect to the periodic orbit, which would have been a good assumption if the particular solution



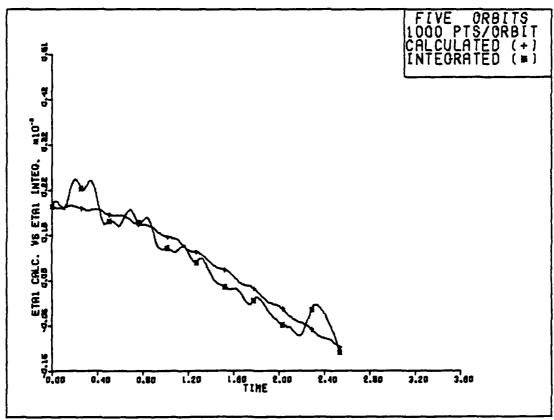
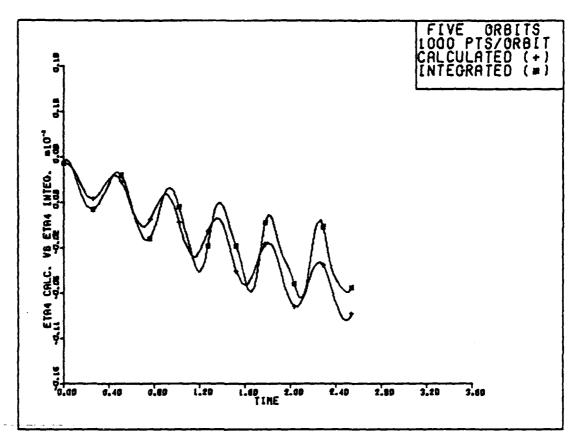


Figure 3(a) ETA vs Time



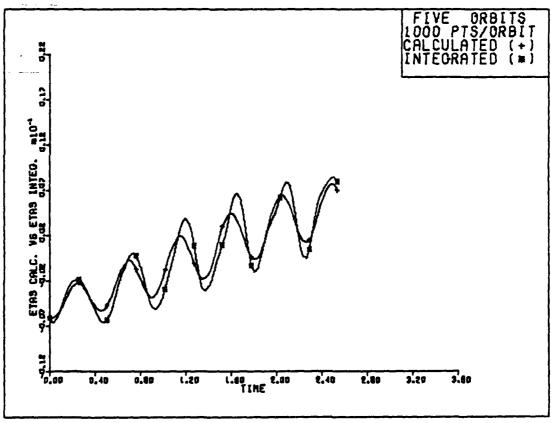
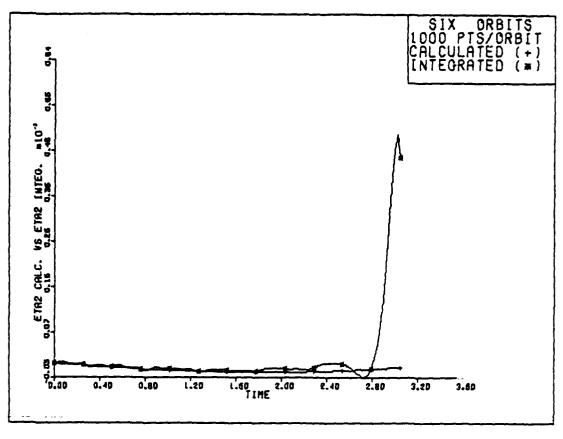
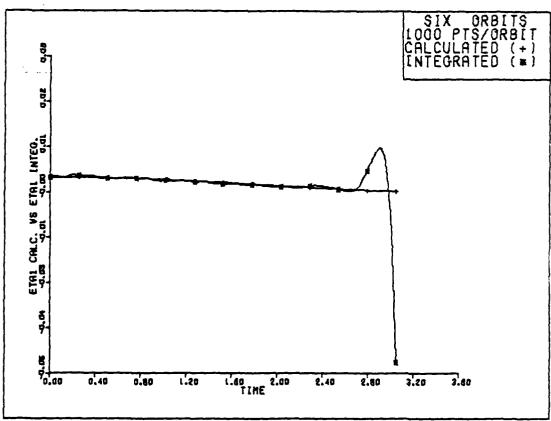
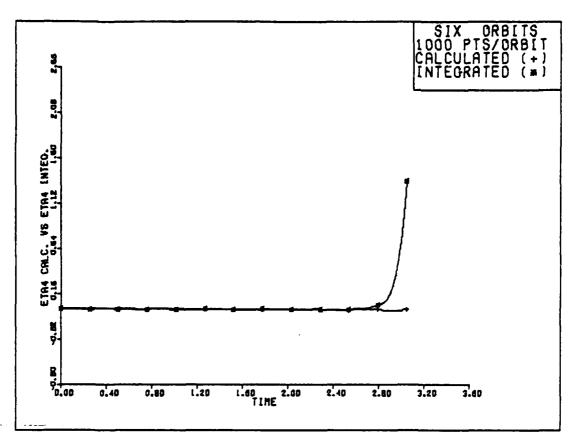


Figure 3(b) ETA vs Time





Ftgure 4(a) ETA vs Time



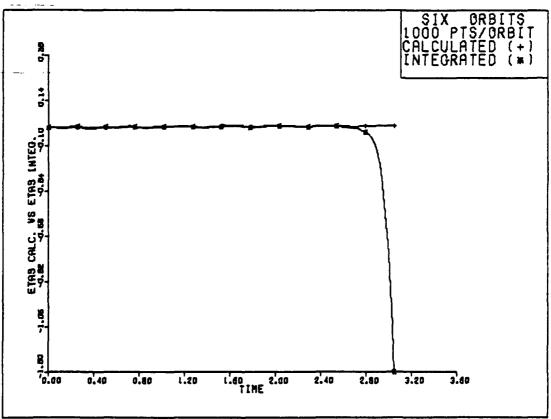


Figure 4(b) ETA vs Time

were two orders of magnitude smaller. A phase portrait for the Shelton Controller with a gain of 0.5 is depicted in Figure 5 (Ref 6). As can be seen in this figure some of the trajectories, on the order of $n_3 = n_d = 10^{-3}$, do not converge to the origin. The combination of factors mentioned above put this solution on one of the trajectories that is unstable.

Due to the combination of factors described previously the system is not stable beyond six orbits. This situation could possibly have been avoided if the particular solution, $\eta_{p3}(t)$, used in the controller was more accurate or if a larger gain was selected for the original controller at the start of the analysis. Selecting a larger gain would have increased the stable operating region of the controller. However, there was no way of actually knowing ahead of time how large or accurate the particular solution would be.

In spite of the failure of the controller to operate beyond six orbits, it did exhibit excellent control characteristics up to the time it became unstable. In Figure 6 we can see that the integrated and calculated solutions track very well. In this case the integrated solution was started on the calculated solution. To see if the controller would drive the integrated solution to the calculated solution, I started the integrated solution off of the calculated solution and the controller did drive it back to the calculated solution (Figure 7). Figure 8 shows the difference between the calculated and integrated solution for five orbits and

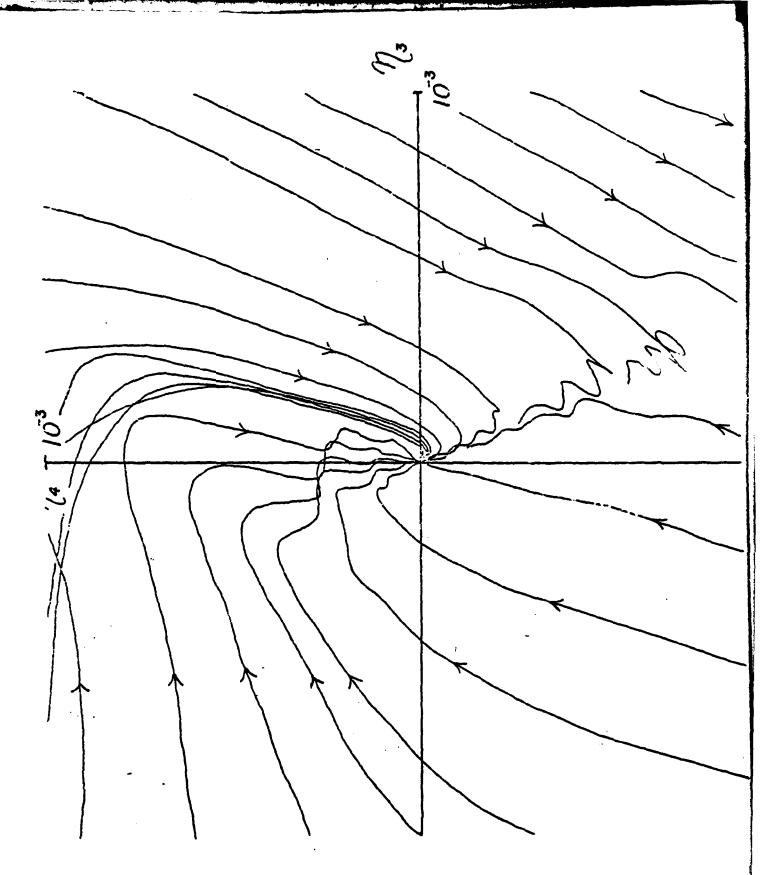


Figure 5 ETA3-4 Phase Portrait

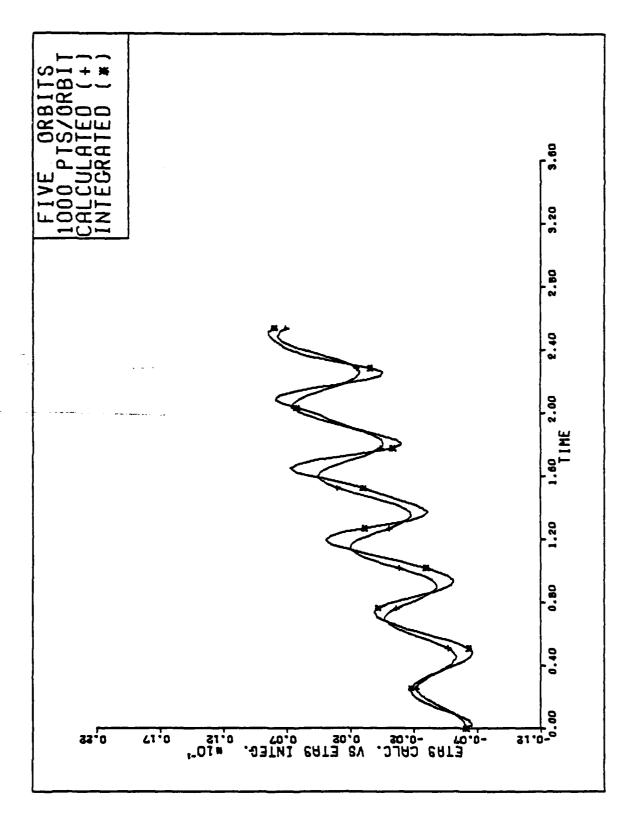


Figure 6 ETA3 Started on Orbit

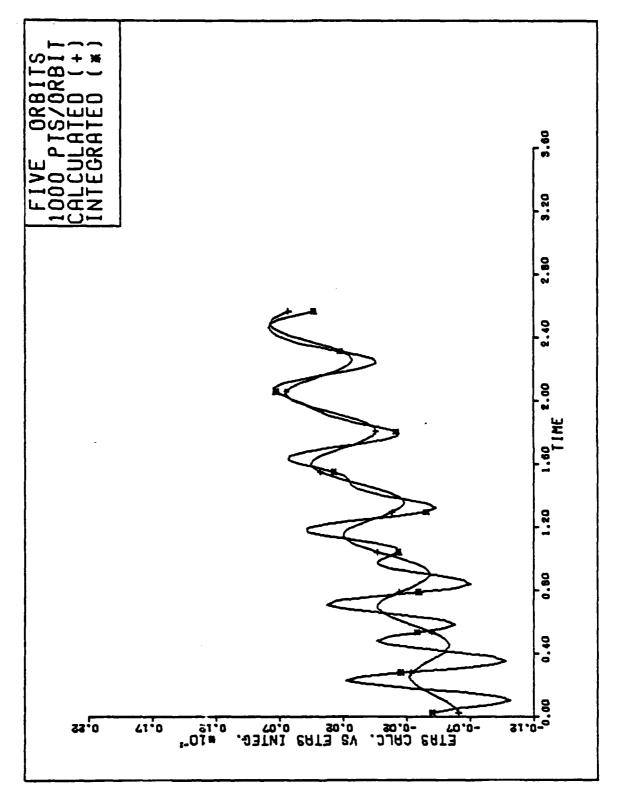
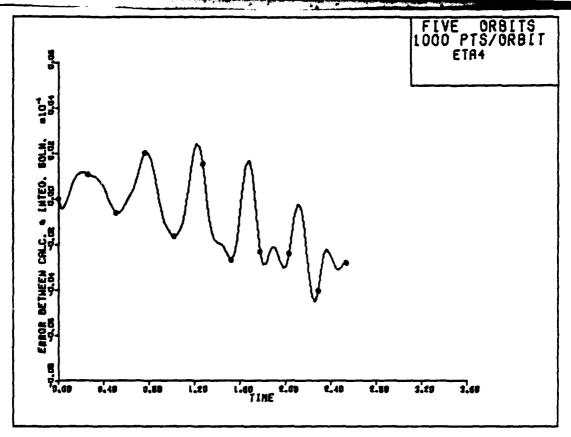


Figure 7 ETA3 Started off Orbit



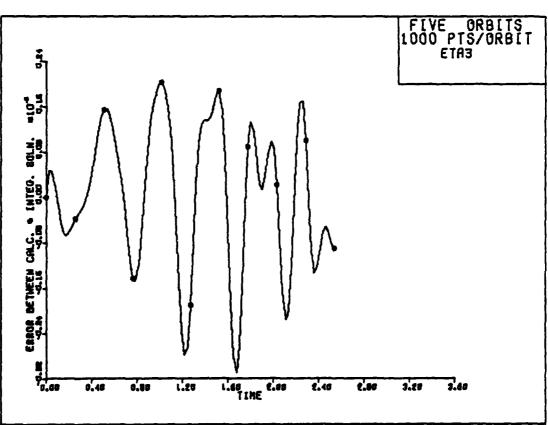
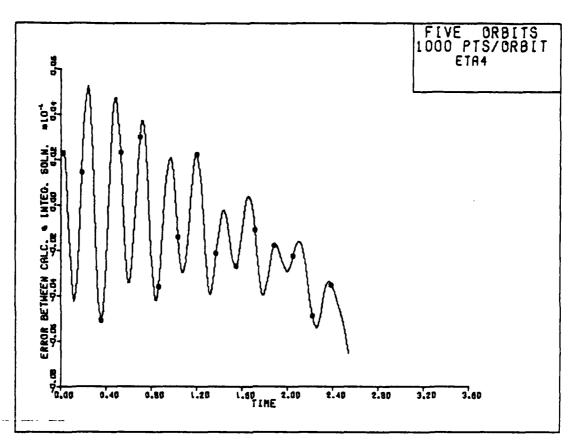


Figure 8(a) Difference Between Calculated and Integrated Solutions



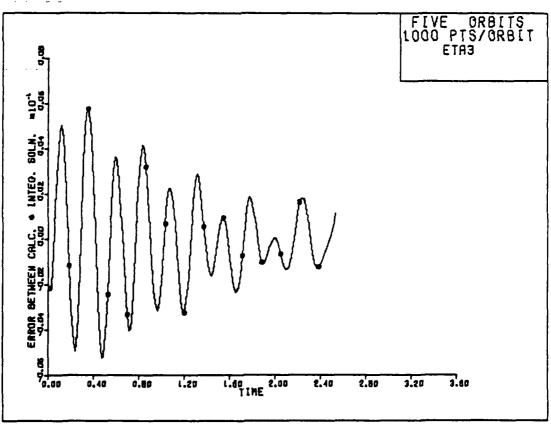


Figure 8(b) Difference Between Calculated and Integrated Solutions

Figure 9 shows the amount of control being expended to maintain the orbit. The integrated control gains (ICG) shown in Figure 9 are defined as

$$ICG(t) = \frac{1}{t} \int control \ acc. | \ d\gamma$$

which is equal to the average control acceleration required for system operation. Expressing these values in terms of g's, yields an average control cost of approximately 10^{-7} g which is comparable to a typical earth-synchronous satellite's control cost of 10^{-7} g (Ref. 2:63).

Conclusions

As shown in Figures 7, 2 and 9 the modfied controller does work very well over this five orbit period. Because the initial selection of gain for the Shelton controller was too low I did not achieve the long term, low cost solution I was expecting to develop. However, because of the initial well behaved aspect of the controller, I believe the method is sound. In order to fully develop this control system and prove that it is feasible, I recommend that a follow up study be done using a higher gain and also include the effects of the inclination of the moon's orbit.

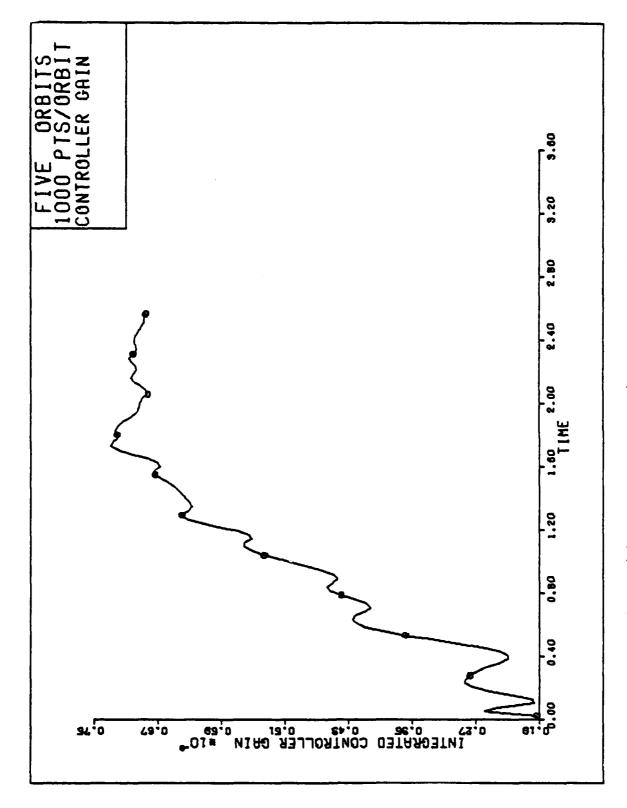


Figure 9(a) Controlled System (Started on Crbit)

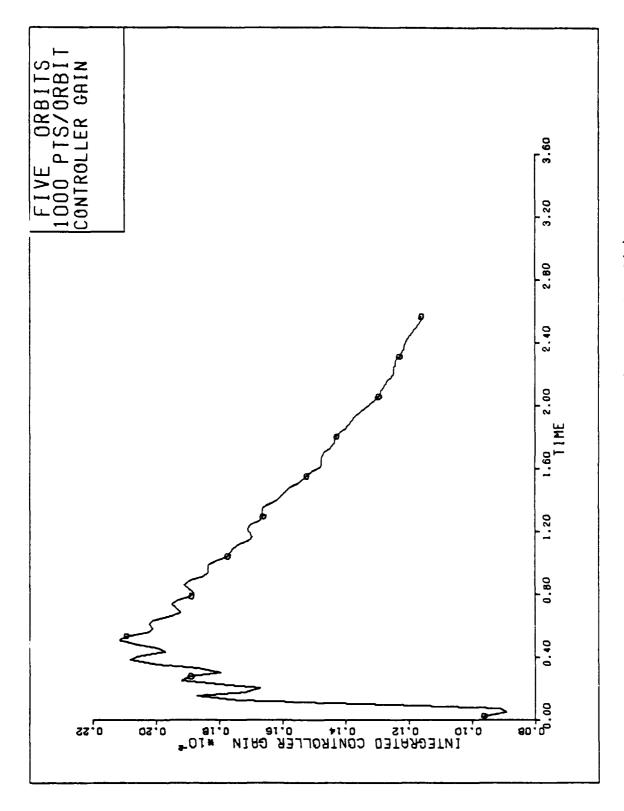


Figure 9(b) Controlled System (Started off Orbit)

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Appendix A

Derivation of the Particular Solution

In deriving the particular solution to (62) we can see that the form of the solutions $\eta_{3p}(t)$ and $\eta_{4p}(t)$ are the same except for a change of indices. If I change the 3's to 4's in (64) on page 24, and follow through the rest of the derivation on pages 24 through 26, the solutions are generically the same.

The differential equations involving $\eta_{\mbox{\footnotesize 1p}}(t)$ and $\eta_{\mbox{\footnotesize 2p}}(t)$ are coupled and of the form

$$\eta_{1p}(t) = \alpha_{1}\eta_{2p}(t) + \left[\Lambda'(t)\overline{P}(t)\right]_{1}$$
(82)

$$\eta_{2p}(t) = \alpha_{2}\eta_{1p}(t) + \left[\Lambda'(t)\overline{P}(t)\right]_{2}$$
(83)

Taking the time derivative of (82) and substituting (83) produces

$$\eta_{1p}(t) = \alpha_1 \alpha_2 \eta_{1p}(t) + \alpha_1 F_2 + F_1$$
 (84)

where \mathbf{F}_1 is the time derivative of $\lceil \Lambda'(t) \bar{P}(t) \rceil_1$ and

$$F_2 = \int \Lambda'(t) \overline{P}(t) 1_2$$
 (85)

Assuming a solution to (84) of the form

$$\eta_{1p}(t) = \sum_{L=1}^{NM} [W_L \cos(\phi_L) + V_L \sin(\phi_L) + Y_L \cos(\psi_L) + Z_L \sin(\psi_L)]$$
 (86)

taking time derivatives of (86) yields

.. NM .2 .2 .2 .2 .2 .2
$$\eta_{lp}(t) = \sum_{L=1}^{\Gamma-W} -W_L \phi_L \cos(\phi_L) - V_L \phi_L \sin(\phi_L) - Y_L \psi_L \cos(\psi_L) - Z_L \psi_L \sin(\psi_L)$$
(88)

with

$$\phi = (nw_C + mw_E)t + mQ_{1C}$$
 (89)

$$\psi = (nw_0 + mw_E)t - mO_{10}$$
 (90)

$$\phi = nw_0 + mw_E \tag{91}$$

$$\dot{\psi} = n w_0 - m w_E \tag{92}$$

$$\phi = \psi = 0 \tag{93}$$

taking the time derivative of $\lceil \Lambda^{\bullet}(t) \overline{P}(t) \rceil_1$, from (61) page 24, yields

$$F_{1} = \frac{1}{2} \sum_{n=0}^{N-1} \sum_{m=0}^{M-1} (A_{nm1} + C_{nm1}) \sin(\psi_{nm}) - \phi_{nm} (A_{nm1} - C_{nm1}) \sin(\phi_{nm})$$

+
$$\psi_{nm} (B_{nm1} - D_{nm1}) \cos (\psi_{nm}) + \phi_{nm} (B_{nm1} + D_{nm1}) \cos (\phi_{nm})$$
 (94)

with $f \Lambda^{\bullet}(t) \overline{P}(t) \Big|_{2}$ equal to

$$F_{2} = \frac{1}{2} \sum_{n=0}^{N-1} \sum_{m=0}^{M-1} (A_{nm2} + C_{nm2}) \cos(\psi_{nm}) + (A_{nm2} - C_{nm2}) \cos(\phi_{nm})$$

+
$$(B_{nm2}-D_{nm2})\sin(\psi_{nm})+(B_{nm2}+D_{nm2})\sin(\phi_{nm})$$
] (95)

taking (87) through (95), substituting them into (84) and equating like terms produces

$$sin(\phi_L)$$
:

$$-V_{L}(nw_{0}+mw_{E})^{2} = \alpha_{1}\alpha_{2}V_{L} + \frac{1}{2}\alpha_{1}(B_{nm2}+D_{nm2}) - (nw_{0}+mw_{E})(A_{nm1}-C_{nm1})$$
(95)

$$cos(\phi_L)$$
:

$$-W_{L}(nw_{0}+mw_{E})^{2} = \alpha_{1}\alpha_{2}W_{L}+\frac{1}{2}\alpha_{1}(A_{nm2}-C_{nm2})+(nw_{0}+mw_{E})(B_{nm1}+D_{nm1})$$
(97)

$$\sin(\psi_L)$$
:

$$-z_{L}(nw_{0}+mw_{E})^{2} = \alpha_{1}\alpha_{2}z_{L}+\frac{1}{2}\alpha_{1}(B_{nm2}-D_{nm2})-(nw_{0}-mw_{E})(A_{nm1}+C_{nm1})$$
(98)

$$-Y_{L}(nw_{0}-mw_{E})^{2} = \alpha_{1}\alpha_{2}Y_{L} + \frac{1}{2}\alpha_{1}(A_{nm2}+C_{nm2}) + (nw_{0}-mw_{E})(B_{nm1}-D_{nm1})$$
(99)

solving for the coefficients of the particular solution from (95) through (99) we obtain

$$W_{1L} = -\frac{\frac{1/2 \alpha_1 (A_{nm2} - C_{nm2}) + 1/2 (nw_0 + mw_E) (B_{nm1} + D_{nm1})}{[\alpha_1 \alpha_2 + (nw_0 + mw_E)^2]}$$
(100)

$$V_{1L} = -\frac{\frac{1/2 \alpha_1 (B_{nm2} + D_{nm2}) - \frac{1}{2} (nw_0 + mw_E) (A_{nm1} - C_{nm1})}{(\alpha_1 \alpha_2 + (nw_0 + mw_E)^2)}$$
(101)

$$Y_{1L} = -\frac{\frac{1/2 \alpha_1 (A_{nm2} + C_{nm2}) + \frac{1}{2} (nw_0 - mw_E) (B_{nm1} - D_{nm1})}{(\alpha_1 \alpha_2 + (nw_0 - mw_E)^2)}$$
(102)

$$Z_{1L} = -\frac{\frac{1/2 \alpha_1 (B_{nm2} - D_{nm2}) - \frac{1}{2} (nw_0 + mw_E) (A_{nm1} + C_{nm1})}{(\alpha_1 \alpha_2 + (nw_0 - mw_E)^2)}$$
(103)

Numerically evaluating (100) through (103) and using them with (86) gives us a solution to $\eta_{1p}(t)$. Equations (100) through (103) can be used to provide the coefficients for $\eta_{2p}(t)$ by switching the sub-indexes, 1 and 2.

Dennis W. Ehrler was born May 15, 1955 in Shreveport, Louisiana. In December, 1961, he and his family moved to a farm near Turtle Lake, Wisconsin where he attended the Turtle Lake school system. In May, 1973, he was graduated from Turtle Lake High school and entered into the U.S. Air Force Academy in June, 1973. He was graduated from the USAFA with a Bachelor of Science degree in Astronautical Engineering in 1977.

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| AFIT/GA/AA/81D-4 AT-14111107 | |
| 4. TITLE (and Subtitle) | 5. TYPE OF REPORT & PERIOD COVERED |
| A Perturbation Analysis of Modal Control | MS Thesis |
| of an Unstable Periodic Orbit | 6. PERFORMING ORG, REPORT NUMBER |
| 7. AUTHOR(s) | 8. CONTRACT OR GRANT NUMBER(#) |
| Dennis W. Ehrler | } |
| Captain, USAF | |
| 9. PERFORMING ORGANIZATION NAME AND ADDRESS | 10. PROGRAM ELEMENT, PROJECT, TASK |
| 3. PERFORMING ORGANIZATION NAME AND ADDRESS | AREA & WORK UNIT NUMBERS |
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| | |
| 11. CONTROLLING OFFICE NAME AND ADDRESS | 12. REPORT DATE |
| | December, 1981 |
| | 13. NUMBER OF PAGES |
| | 60 |
| 14. MONITORING AGENCY NAME & ADDRESS(if different from Controlling Office) | 15. SECURITY CLASS. (of this report) |
| | 1 |
| | 15a. DECLASSIFICATION DOWNGRADING |
| | SCHEDULE |
| 16. DISTRIBUTION STATEMENT (of this Report) | |
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| 17. DISTRIBUTION STATEMENT (of the abstract entered in Block 20, If different from Report) | |
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| 18. SUPPLEMENTARY NOTES | |
| Approved for public release, IAW AFR | 190-17 |
| Frederic C. Lynch, Major, USAF | $4 \cdot 1 \cdot 0 \cdot 1 \cdot 1$ |
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| eccentricity on a satellite in orbit about L3 was costructed and verified | |

A Fourier representation of the perturbation effects of the moon's eccentricity on a satellite in orbit about L3 was costructed and verified numerically. It was then incorporated in a modal control scheme for linear systems that are time periodic which was developed by Shelton. The feasibility of this modified control scheme was then verified by computer simulation by controlling the actual non-linear motions of the satellite with the modified controller.

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